

## SYSTEMS AND METHODS FOR BASE STATION SIGNAL TRANSMISSION POWER CONTROL

### FIELD OF THE INVENTION

5       The present invention relates generally to a wireless radio network and, more particularly, to systems and methods for controlling the power of a Base Station (BS) transmitter in the wireless radio network.

### BACKGROUND OF THE INVENTION

10       In a wireless radio network, a Base Station (BS) communicates with a Mobile Station (MS), e.g., a cellular phone. One exemplary wireless radio network is a Code Division Multiple Access (CDMA) cellular communications system. In a forward direction (downlink) of the CDMA system, the BS transmits a signal, which is divided into 20 ms frames, to the MS over a forward traffic channel. Typically, the CDMA network employs a power control system to control the transmission power of the BS transmitter in order to maintain a desired Frame Error Rate (FER) of the received signal at the MS.

15       A typical power control system comprises an outer loop power control and an inner loop power control, both of which are located in the MS. The outer loop power control compares an estimation of the FER of the received signal with a target FER, typically one percent. The outer loop power control then adjusts an  $E_b/N_t$  (energy per bit to total noise density) set point based on this comparison, where  $E_b/N_t$  is a measure of the quality of the received signal. If the estimated FER is greater than the target FER, then the outer loop power control increases the  $E_b/N_t$  set point by a certain amount. Otherwise, the outer loop power control decreases the  $E_b/N_t$  set point by a certain amount.

The inner loop power control receives the  $E_b/N_t$  set point from the outer loop power control. The inner loop power control compares the  $E_b/N_t$  set point with an estimation of the  $E_b/N_t$  of the received signal. The inner loop power control then outputs a power control command based on this comparison. If the estimated  $E_b/N_t$  is less than the  $E_b/N_t$  set point, then the inner loop power control outputs a power control command instructing the BS transmitter to increase its transmission power. Otherwise, the inner loop power control outputs a power control command instructing the BS transmitter to decrease its transmission power.

The MS transmits the power control command to the BS over a reverse (uplink) channel. The BS transmitter then increases or decreases its transmission power by a certain amount, e.g., 1 dB, based on the received power control command from the MS.

A drawback of this power control system is that the outer loop power loop typically increases the  $E_b/N_t$  set point by a constant amount when the estimated FER is greater than the target FER, regardless of the number of bit errors within each frame of the received signal. This is because the estimated FER only provides information about the number of frames that are received in error by the MS and not the number bit errors within each received frame.

Therefore, there is a need for a power control system comprising an outer loop power control that is able to make fine adjustments to the  $E_b/N_t$  set point based on additional information about the received frames. This would provide higher performance and accuracy in adjusting the  $E_b/N_t$  set point to achieve a desired FER of the received signal at the MS.

## SUMMARY OF THE INVENTION

The present invention addresses the aforementioned problems by providing a power control system comprising an outer loop power control that is able to make fine adjustments to the  $E_b/N_t$  set point based on additional information about the frames received by the MS.

5 In one embodiment of the present invention, the BS transmitter transmits a signal to the MS over a forward traffic channel. The signal is divided into frames, and each frame is further divided into Logical Transmission Units (LTU). Each LTU comprises an LTU Cyclic Red Check (CRC) field that enables the MS to individually detect an error in each LTU of a received frame.

10 The MS comprises an MS receiver, an FER estimator, an outer loop power control, an inner loop power control and an MS transmitter. The MS receiver demodulates and decodes the signal from the BS. The MS receiver also calculates the number of LTU errors in a currently received frame of the received signal by checking the LTU CRC field in each LTU of the currently received frame. The FER estimator estimates a FER of the decoded signal from the MS receiver and outputs an estimated FER.

15 The outer loop control receives the estimated FER from the FER estimator and the number of LTU errors in the currently received frame from the MS receiver. The outer loop power control then compares the estimated FER from the FER estimator with a target FER, e.g., one percent. If the estimated FER is greater than the target FER, then the outer loop power control increases the  $E_b/N_t$  set point by an up step value that is a function of the number of LTU errors received from the MS  
20 receiver. Otherwise, the outer loop power control decreases the  $E_b/N_t$  set point by a down step value. In one variation of the invention, the down step value may be varied according to the number of LTU errors.

The inner loop power control receives the  $E_b/N_t$  set point from the outer loop power control and an estimated  $E_b/N_t$  of the received signal from the MS receiver. The inner loop power control then compares the  $E_b/N_t$  set point with the estimated  $E_b/N_t$  from the MS receiver. If the estimated  $E_b/N_t$  is less than the  $E_b/N_t$  set point, then the inner loop power control outputs a power control command instructing the BS transmitter to increase its transmission power. Otherwise, the inner loop power control outputs a power control command instructing the BS transmitter to decrease its transmission power. The MS transmitter then transmits the power control command to the BS over a reverse pilot channel.

An advantage of the outer loop power control according to the present invention is that it is able to make finer adjustments to the  $E_b/N_t$  set point compared with the prior art. This is because the outer loop power control of the prior art increases the  $E_b/N_t$  set point by a constant amount when the estimated FER is greater than the target FER, regardless of the number of bit errors within each frame of the received signal. The outer loop power control according to the present invention, on the other hand, varies the amount that the  $E_b/N_t$  set point is increased based on the number of LTU errors of a currently received frame of the received signal. This results in a greater range in the magnitude of adjustments possible for fine accuracy in the adjustment of the  $E_b/N_t$  set point.

In another embodiment of the invention, the FER estimator is replaced with an LTU error rate estimator. The LTU error rate estimator estimates an LTU error rate of the received signal by checking the LTU CRC fields in the received signal. The outer loop receives the estimated LTU error rate from the LTU error rate estimator and a target LTU error rate. The outer loop power control then adjusts the  $E_b/N_t$  set point based on a comparison of the estimated LTU error rate with the target LTU error rate. If the estimated LTU error rate is greater than the target LTU error rate,

then the outer loop power control increases the  $E_b/N_t$  set point by a certain amount. Otherwise, the outer loop power control decreases the  $E_b/N_t$  set point by a certain amount.

Other objects and features of the present invention will become apparent from consideration of the following description taken in conjunction with the accompanying drawings.

## BRIEF DESCRIPTION OF THE DRAWINGS

The drawings illustrate both the design and utility of the preferred embodiments of the present invention, in which similar elements in different embodiments are referred to by the same reference numbers for purposes of ease in illustration of the invention, wherein:

5        FIG. 1 is a block diagram of an exemplary power control system according to an embodiment of the present invention.

FIG. 2 is a diagram of a frame structure according to an embodiment of the present invention.

FIG. 3 is a block diagram of an exemplary power control system according to another embodiment of the present invention.

## DETAILED DESCRIPTION OF THE PREFERRED EMBODIMENTS

FIG. 1 shows an exemplary power control system 8 according to an embodiment of the present invention. The power control system 8 comprises a Base Station BS 15 and a Mobile Station (MS) 20, e.g., a cellular phone. The BS 15 further comprises a BS transmitter 25 having a power control input 28 and a BS receiver 40 coupled to the power control input 28 of the BS transmitter 25. The MS 20 further comprises an MS receiver 45, a frame error rate (FER) estimator 50 coupled to the MS receiver 45 and an outer loop power control 60 coupled to an output 55 of the FER estimator 50. The MS 20 also comprises an inner loop power 70 control coupled to an output 65 of the outer loop power control 60 and an MS transmitter 80 coupled to an output 75 of the inner control power loop 70.

During operation, the BS transmitter 25 modulates an incoming signal 22 into a spread-spectrum signal 30 using a spreading code. The BS transmitter 25 then transmits the spread-spectrum signal 30 using a forward traffic channel 35 to the MS 20. Preferably, the forward traffic channel 35 comprises at least one Supplemental Channel (SCH) that carries non-voice data to the MS 20. The non-voice data may be used for providing fax and Internet applications to the MS 20. The forward traffic channel 35 may also comprise a fundamental channel (FCH) that carries voice data to the MS 20.

The present invention preferably uses an SCH according to the IS-2000 standard established by the Telecommunications Industry Association (TIA). The SCH according to the IS-2000 standard is divided into 20 ms SCH frames. FIG. 2 shows an exemplary SCH frame structure according to the IS-2000 standard. The SCH frame 210 comprises a physical layer Cyclic Redundancy Check (CRC) 220, which is typically 16 bits in length. The physical layer CRC 220 enables the MS

20 to detect a frame error of a received SCH frame 210 by inputting the physical layer CRC 220 into a state machine that outputs all '0's when the SCH frame 210 is correctly received.

The SCH frame 210 further comprises an information data field 230 containing Multiplex layer Protocol Data Units (MuxPDU). The information data field 230 is divided into Link  
5 Transmission Units (LTUs) 240-n. The information data field 230 may be divided into 2, 4 or 8 LTUs 240-n. Each LUTU 240-n contains one or more MuxPDUs, depending on the chosen multiplex option provided by the IS-2000 standard. In FIG. 2, each LUTU 240-n may contain either a 1 double size MuxPDU or 2 single size MuxPDUs 245. Each LUTU 240-n further includes a LUTU CRC field 250, which is 16 bits in length. The LUTU CRC field 250 in each LUTU 240-n enables the MS 20 to individually detect an error in each LUTU 240-n of a received SCH frame 210. The SCH frame 210 may also comprise a field 260 containing a variable number of '0's and an encoder tail field 265, which is 8 bits in length.

At the MS 20, the MS receiver 45 demodulates the received spread-spectrum signal 38 by despreading the spread-spectrum signal 38. The MS receiver 45 then decodes the demodulated signal and outputs the decoded signal 48 to the FER estimator 50. The FER estimator 50 estimates a  
10 Frame Error Rate (FER) of the decoded signal 48, where the FER represents the percentage of frames of the decoded signal 48 that are received in error by the MS 20. To estimate the FER, the FER estimator 50 detects the number of frame errors of the decoded signal 48 within a time period by checking the physical layer CRC 250 of each frame within the time period. The FER estimator  
20 50 may then employ any one or a combination of well known techniques in the art for estimating a FER based on the number of detected frame errors of a decoded signal.

The outer loop power control 60 receives the estimated FER 55 from the FER estimator 50 at a rate of about one estimated FER per frame of the decoded signal 48. The outer loop power control



60 also receives a target FER 62, e.g., one percent. In addition, the outer loop power control 60 receives a number of LTU errors 68, denoted  $N_E$ , in a currently received frame of the decoded signal 48 from the MS receiver 45. The MS receiver 45 calculates the number of LTU errors,  $N_E$ , in the currently received frame by checking the LTU CRC field 240-n of each LTU in the currently received frame.

The outer loop power control 60 uses the estimated FER 55, the target FER 62 and the number of LTU errors 68,  $N_E$ , in the currently received frame to adjust an  $E_b/N_t$  (energy per bit to total noise density) set point, where  $E_b/N_t$  is a measure of the quality of the received signal 38. To adjust the current  $E_b/N_t$  set point,  $E_b/N_{t,CURR}$ , the outer loop power control 60 compares the currently received estimated FER 55 from the FER estimator 50 with the target FER 62.

If the estimated FER 55 is higher than the target FER 62, then the outer loop power control 60 increases the previous  $E_b/N_t$  set point,  $E_b/N_{t,PREV}$ , according to the formula:

$$E_b/N_{t,CURR} = E_b/N_{t,PREV} + \Delta_{UP}(N_E)$$

where  $E_b/N_{t,PREV}$  is increased by an up step value of  $\Delta_{UP}(N_E)$ , the magnitude of which is a function of the number of LTU errors 68,  $N_E$ , from the MS receiver 45. The up step value  $\Delta_{UP}(N_E)$  for each possible number of LTU errors,  $N_E$ , may be provided to the outer loop power control 60 by a lookup table that assigns an up step value to each possible  $N_E$ . Preferably, the magnitude of the up step value  $\Delta_{UP}(N_E)$  increases for increasing  $N_E$ . This is because an increasing  $N_E$  is usually caused by worsening channel conditions between the BS 15 and the MS 20 in the forward direction.

If the estimated FER 55 is less than the target FER 62, then the outer loop power control 60 decreases the previous  $E_b/N_t$  set point,  $E_b/N_{t,PREV}$ , according to the formula:

$$E_b/N_{t,CURR} = E_b/N_{t,PREV} - \Delta_{DOWN}$$

where  $E_b/N_{t_{\text{PREV}}}$  is decreased by a down step value of  $\Delta_{\text{Down}}$ .

The outer loop power control 60 adjusts the  $E_b/N_t$  set point each time it receives an estimated FER 55 from the FER estimator 50, which occurs once per frame of the decoded signal. As a result, the outer loop power control 60 provides frame-by-frame adjustment of the  $E_b/N_t$  set point.

5       The inner loop power control receives 70 the  $E_b/N_t$  set point 65 from the outer loop power control 60 at a rate of about once per frame. In addition, the inner loop power control 70 receives an estimated  $E_b/N_t$  72 of the received signal 38 from the MS receiver 45. The MS receiver 45 outputs the estimated  $E_b/N_t$  72 to the inner loop power control 70 at a rate greater than once per frame, preferably 16 times per frame. The inner loop power control 70 compares the received  $E_b/N_t$  set point 65 with the estimated  $E_b/N_t$  72 from the MS receiver 45. The inner loop power control 70 then outputs a power control command 75 based on the comparison. If the estimated  $E_b/N_t$  is less than the  $E_b/N_t$  set point, then the inner loop power control 70 outputs a power control command 75 instructing the BS 15 to increase its transmission power. Otherwise, the inner loop power control 70 outputs a power control command 75 instructing the BS 15 to decrease its transmission power. The power command may be one bit in length in which a bit value of one instructs the BS 15 to increase its transmission power and a bit value of zero instructs the BS 15 to decrease its transmission power. The inner loop power control 70 outputs a power control command 75 each time that it receives an estimated  $E_b/N_t$  72 from the MS receiver 45. Preferably, the inner loop power control 70 outputs a power control command 70 at a rate of about 16 times per frame. For a frame length of about 20ms,  
20   the inner loop power control 70 outputs a power control command 75 every 1.25 ms.

The MS transmitter 80 receives power control commands 75 from the inner loop power control 70 and a pilot signal 82, which may be a pseudo random sequence. The pilot signal 82 is divided into 20 ms frames. Each frame of the pilot signal is further divided into 16 power control

groups (PCGs) with each PCG having a length of 1.25 ms. The MS transmitter 80 multiplexes the power control commands 75 with the pilot signal 82 so that one power control command is placed into each PCG of the pilot signal 82. The MS transmitter 80 then transmits the pilot signal 85 multiplexed with the power control commands to the BS 15 using a reverse pilot channel 90.

5       The BS receiver 40 demodulates and decodes the received pilot signal 90 from the MS 20, and extracts the power control commands from the pilot signal 90. The extracted power control commands are sent to the power control input 28 of the BS transmitter 25. The BS transmitter 25 then adjusts its transmission power level by a predetermined amount, e.g., 1 dB, based on the received power control commands. The BS transmitter 25 may adjust its transmission power level by adjusting the power level of a power amplifier (not shown) in the BS transmitter 25.

10       An advantage of the power control system 8 of the present invention is that the outer loop power control 60 is able to make finer adjustments to the  $E_b/N_t$  set point compared with the prior art. This is because the power control system of the prior art increases the  $E_b/N_t$  set point by a constant amount when the estimated FER is greater than the target FER, regardless of the number of bit errors within each received frame. The power control system 8 of the present invention, on the other hand, varies the amount that the  $E_b/N_t$  set point is increased based on the number of LTU errors in a currently received frame. This result in greater granularity in the adjustment of the  $E_b/N_t$  set point so as to improve the sensitivity of base station power adjustments. In addition, the number of LTU errors in the currently received frame provides the outer loop power control 70 with greater  
20       resolution in assessing the channel conditions between the BS 15 and the MS 20 than the case where only the estimated FER is used.

      The up step value  $\Delta_{UP}(N_E)$  for each possible number of LTU errors,  $N_E$ , in a received frame may be found experimentally or through a computer simulation. This may be done, for example, by

adjusting the up step value  $\Delta_{UP}(N_E)$  for each number of LTU errors,  $N_E$ , to a value that minimizes both a measured FER and a measured  $E_b/N_t$ . This may help to further optimize the forward link channel capacity and the power efficiency of the BS 15.

In another embodiment of the invention, the outer loop power control 60 also varies the down step value  $\Delta_{DOWN}$  as a function of the number of LTU errors 68 from the MS receiver 45. In this embodiment, when the estimated FER 55 is less than the target FER 62, the outer loop adjusts the current  $E_b/N_t$  set point,  $E_b/N_{t,CURR}$ , according to the formula:

$$E_b/N_{t,CURR} = E_b/N_{t,PREV} - \Delta_{DOWN}(N_E)$$

where the down step value  $\Delta_{DOWN}(N_E)$  is now a function of the number of LTU errors,  $N_E$ , of the currently received frame. Preferably, the magnitude of the down step value  $\Delta_{DOWN}(N_E)$  decreases with increasing  $N_E$ . The down step value  $\Delta_{DOWN}(N_E)$  for each possible  $N_E$  may be found by employing methods similar to those used to find the up step value  $\Delta_{UP}(N_E)$  for each possible  $N_E$ .

Even though the outer loop power control 60 adjusts the  $E_b/N_t$  set point once per frame, preferably, the outer loop power control 60 does not increase the  $E_b/N_t$  set point twice when two consecutive estimated FERs 55 from the FER estimator 50 are greater than target FER 62. This is because there is typically a time delay before a power control command based on an  $E_b/N_t$  set point is received by the BS transmitter 15. Because of this time delay, the outer loop power control 70 can not immediately determine if an increase in the  $E_b/N_t$  set point is sufficient to reduce the estimated FER 55 below the target FER 62 or if an additional increase in the  $E_b/N_t$  set point is necessary.

Thus, in a preferred embodiment, the outer loop power control 70 waits for a time period of about 3 to 4 frames after increasing the  $E_b/N_t$  set point before increasing the  $E_b/N_t$  set point again, even when the estimated FER 55 is greater than the target FER 62 within the time period. The time

period is chosen to give the outer loop control power 60 enough time to determine the effect of an increase in the  $E_b/N_t$  set point on the estimated FER 55. Within this time period, the outer loop control 70 decreases the  $E_b/N_t$  by the step value  $\Delta_{\text{DOWN}}$  for each frame.

FIG. 3 shows a power control system 310 according to another embodiment of the invention.

5 In this embodiment, a LTU error rate estimator 320 replaces the FER estimator 50 of FIG. 1. The LTU error rate estimator 320 estimates a LTU error rate for the decoded signal 48, i.e. the percentage of LTUs in the decoded signal 48 that are received in error by the MS 20. The LTU error rate estimator 320 estimates the LTU error rate by checking the LTU CRC fields in the LTUs of the decoded signal 48.

10 The outer loop power control 330 receives the estimated LTU error rate 325 from the LTU error rate estimator 320, preferably once per frame. The outer loop power control 330 also receives a target LTU error rate 335. The outer loop power control 330 then compares the estimated LTU error rate 325 with the target LTU error rate 335. If the estimated LTU error rate is greater than the target LTU error rate 335, then the outer loop power control 330 increases the  $E_b/N_t$  set point 65 by a step value of  $\Delta_{\text{UP}}$ . Otherwise, the outer loop power control 330 decreases the  $E_b/N_t$  set point 65 by a step value of  $\Delta_{\text{DOWN}}$ . In this embodiment, the outer loop power loop 330 does not vary the up step value  $\Delta_{\text{UP}}$  based on the number of LTU errors in a received frame. This is because information based on LTU errors in the decoded signal 48 is already incorporated into the estimated LTU error rate 325 from the LTU error rate estimator 320. Thus, the power control system 310 according to  
20 this embodiment adjusts the transmission power of the BS 15 transmitter to achieve a desired LTU error rate at the MS 20.

While various embodiments of the application have been described, it will be apparent to those of ordinary skill in the art that many more embodiments and implementations are possible that are within the scope of the present invention.

For example, even though  $E_b/N_t$  was used as a measure of the quality of the received signal, those skilled in the art will appreciate that another quantity measure may be used by the outer loop power control 60 for the same purpose. For example, many BS transmitters use an orthogonal Walsh code to modulate an incoming signal. In this case, the outer loop power control loop 60 may adjust an  $E_w/N_t$  (Energy per Walsh code to total noise density) set point instead of an  $E_b/N_t$  set point. The outer loop power control 60 may also receive an estimated  $E_w/N_t$  of the received signal 38 from the MS receiver instead of an estimated  $E_b/N_t$ . In addition, the MS receiver 20 may output an average of the number of LTU errors per received frame instead of outputting the number of LTU errors in the currently received frame. Further, the MS is not limited to a cellular telephone but may be one or more of any number of wireless devices including a personal digital assistant (PDA), web pad, laptop personal computer (PC), etc. Therefore, the invention is not to be restricted or limited except in accordance with the following claims and their equivalents.